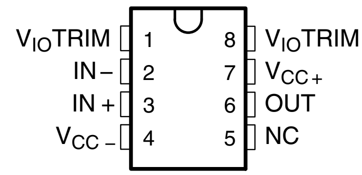
- **Replacements for ADI, PMI and LTC OP27 Series**

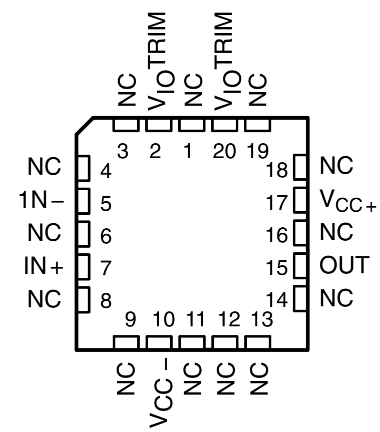
Features of OP27A and OP27C:

- **Maximum Equivalent Input Noise Voltage:**
  3.8 nV/√Hz at 1 kHz
  5.5 nV/√Hz at 10 kHz
- **Very Low Peak-to-Peak Noise Voltage at 0.1 Hz to 10 Hz . . . 80 nV Typ**
- **Low Input Offset Voltage**
  OP27A . . . 25 μV Max
  OP27C . . . 100 μV Max
- **High Voltage Amplification**
  OP27A . . . 1 V/μV Min
  OP27C . . . 0.7 V/μV Min

JG PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



NC − No internal connection

description

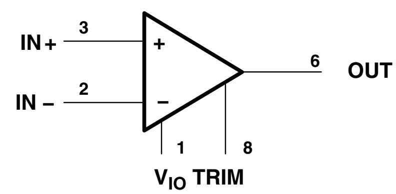
The OP27 operational amplifiers combine outstanding noise performance with excellent precision and high-speed specifications. The wideband noise is only 3 nV/√Hz and with the 1/f noise corner at 2.7 Hz, low noise is maintained for all low-frequency applications.

The outstanding characteristics of the OP27 make these devices excellent choices for low-noise amplifier applications requiring precision performance and reliability.

The OP27 series is compensated for unity gain.

The OP27A and OP27C are characterized for operation over the full military temperature range of −55°C to 125°C.

symbol

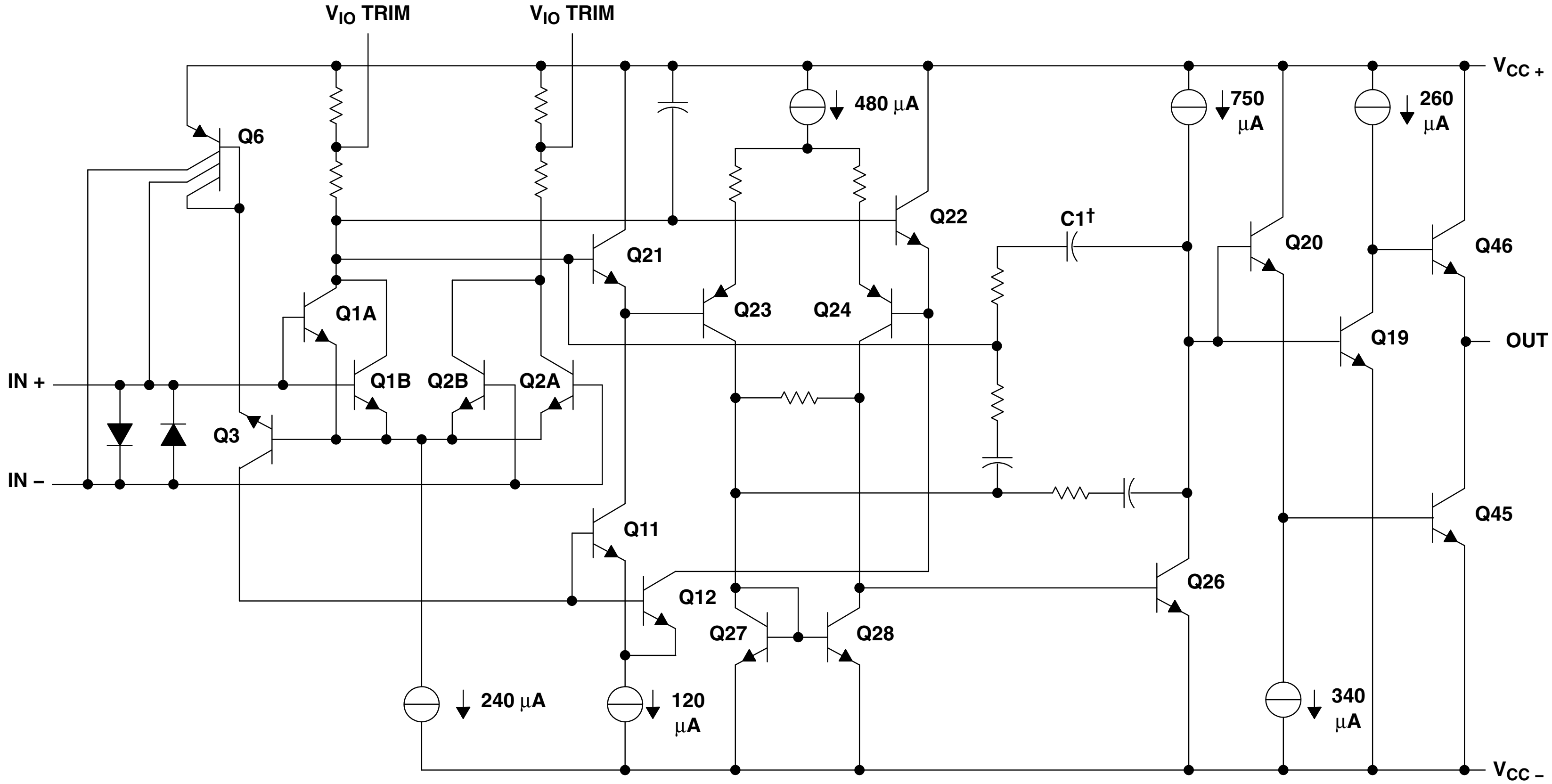


Pin numbers are for the JG packages.

AVAILABLE OPTIONS

T_A	V_{IO}max AT 25°C	STABLE GAIN	PACKAGE	
			CERAMIC DIP (JG)	CHIP CARRIER (FK)
−55°C to 125°C	25 μV	1	OP27AJG	OP27AFK
	100 μV	1	OP27CJG	—

schematic



† C1 = 120 pF for OP27

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V_{CC+} (see Note 1) . . . . 22 V
Supply voltage, V_{CC-} (see Note 1) . . . . 22 V
Input voltage, V_I . . . . $V_{CC\pm}$
Duration of output short circuit . . . . unlimited
Differential input current (see Note 2) . . . . ±25 mA
Continuous power dissipation . . . . See Dissipation Rating Table
Operating free-air temperature range: OP27A, OP27C . . . . –55°C to 125°C
Storage temperature range . . . . –65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG or FK package . . . . 300°C

NOTES: 1. All voltage values are with respect to the midpoint between V_{CC+} and V_{CC-} unless otherwise noted.
2. The inputs are protected by back-to-back diodes. Current-limiting resistors are not used in order to achieve low noise. Excessive input current will flow if a differential input voltage in excess of approximately ±0.7 V is applied between the inputs unless some limiting resistance is used.

DISSIPATION RATING TABLE

PACKAGE	$T_A \le 25°C$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25°C$	$T_A = 85°C$ POWER RATING	$T_A = 125°C$ POWER RATING
JG	1050 mW	8.4 mW/°C	546 mW	210 mW
FK	1375 mW	11.0 mW/°C	715 mW	275 mW

recommended operating conditions

		OP27A MIN	OP27A NOM	OP27A MAX	OP27C MIN	OP27C NOM	OP27C MAX	UNIT
Supply voltage, V_{CC+}		4	15	22	4	15	22	V
Supply voltage, V_{CC-}		−4	−15	−22	−4	−15	−22	V
Common-mode input voltage, V_{IC}	$V_{CC\pm} = \pm 15$ V, $T_A = 25°C$	± 11			±11			V
	$V_{CC\pm} = \pm 15$ V, $T_A = -55°C$ to 125°C	±10.3			±10.2			
Operating free-air temperature, T_A		−55		125	−55		125	°C

electrical characteristics at specified free-air temperature, $V_{CC\pm} = \pm15$ V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A†	OP27A MIN	OP27A TYP	OP27A MAX	OP27C MIN	OP27C TYP	OP27C MAX	UNIT
V_{IO}	Input offset voltage	$V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$, See Note 3	25°C		10	25		30	100	μV
			Full range			60			300	
α_{VIO}	Average temperature coefficient of input offset voltage		Full range		0.2	0.6		0.4	1.8	μV/°C
	Long-term drift of input offset voltage	See Note 4			0.2	1		0.4	2	μV/mo
I_{IO}	Input offset current	$V_O = 0$, $V_{IC} = 0$	25°C		7	35		12	75	nA
			Full range			50			135	
I_{IB}	Input bias current	$V_O = 0$, $V_{IC} = 0$	25°C		±10	±40		±15	±80	nA
			Full range			±60			±150	
V_{ICR}	Common-mode input voltage range		25°C	11 to −11			11 to −11			V
			Full range	10.3 to −10.3			10.5 to −10.5			
V_{OM}	Peak output voltage swing	$R_L \geq 2$ kΩ		±12	±13.8		±11.5	±13.5		V
		$R_L \geq 0.6$ kΩ		±10	±11.5		±10	±11.5		
		$R_L \geq 2$ kΩ	Full range	±11.5			10.5			
A_{VD}	Large-signal differential voltage amplification	$R_L \geq 2$ kΩ, $V_O = \pm10$ V		1000	1800		700	1500		V/mV
		$R_L \geq 1$ kΩ, $V_O = \pm10$ V		800	1500			1500		
		$R_L \geq 0.6$ kΩ, $V_O = \pm1$ V, $V_{CC\pm} = \pm 4$ V		250	700		200	500		
		$R_L \geq 2$ kΩ, $V_O = \pm10$ V	Full range	600			300			
$r_{i(CM)}$	Common-mode input resistance				3			2		GΩ
r_o	Output resistance	$V_O = 0$, $I_O = 0$	25°C		70			70		Ω
CMRR	Common-mode rejection ratio	$V_{IC} = \pm11$ V	25°C	114	126		100	120		dB
		$V_{IC} = \pm10$ V	Full range	110			94			
k_{SVR}	Supply voltage rejection ratio	$V_{CC\pm} = \pm4$ V to ±18 V	25°C	100	120		94	118		dB
		$V_{CC\pm} = \pm4.5$ V to ±18 V	Full range	96			86			

† Full range is – 55°C to 125°C.

NOTES: 3. Input offset voltage measurements are performed by automatic test equipment approximately 0.5 seconds after applying power.

4. Long-term drift of input offset voltage refers to the average trend line of offset voltage versus time over extended periods after the first 30 days of operation. Excluding the initial hour of operation, changes in V_{IO} during the first 30 days are typically 2.5 μV (see Figure 3).

OP27 operating characteristics, $V_{CC\pm} = \pm15$ V, $T_A = 25°C$

PARAMETER		TEST CONDITIONS		OP27A			OP27C			UNIT
				MIN	TYP	MAX	MIN	TYP	MAX	
SR	Slew rate	$A_{VD} \geq 1$,	$R_L \geq 2$ kΩ	1.7	2.8		1.7	2.8		V/μs
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to 10 Hz, See Figure 26	$R_S = 20$ Ω,		0.225	0.375		0.225	0.375	μV
V_n	Equivalent input noise voltage	f = 10 Hz,	$R_S = 20$ Ω		3.5	8		3.8	8	nV/√Hz
		f = 1 kHz,	$R_S = 20$ Ω		3	4		3.2	4	
I_n	Equivalent input noise current	f = 10 Hz,	See Figure 27		5	25		5	25	pA/√Hz
		f = 1 kHz,	See Figure 27		0.7	2.5		0.7	2.5	
	Gain-bandwidth product	f = 100 kHz		5	8		5	8		MHz

TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	vs Temperature	1
ΔV_{IO}	Change in input offset voltage	vs Time after power on vs Time (long-term drift)	2 3
I_{IO}	Input offset current	vs Temperature	4
I_{IB}	Input bias current	vs Temperature	5
V_{ICR}	Common-mode input voltage range	vs Supply voltage	6
V_{OM}	Maximum peak output voltage	vs Load resistance	7
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	8
A_{VD}	Differential voltage amplification	vs Supply voltage vs Load resistance vs Frequency	9 10 11, 12
CMRR	Common-mode rejection ratio	vs Frequency	13
k_{SVR}	Supply voltage rejection ratio	vs Frequency	14
SR	Slew rate	vs Temperature	15
ϕ_m	Phase margin	vs Temperature	16
ϕ	Phase shift	vs Frequency	11
V_n	Equivalent input noise voltage	vs Bandwidth vs Source resistance vs Supply voltage vs Temperature vs Frequency	17 18 19 20 21
	Gain-bandwidth product	vs Temperature	16
I_{OS}	Short-circuit output current	vs Time	22
I_{CC}	Supply current	vs Supply voltage	23
	Pulse response	Small signal Large signal	24 25

TYPICAL CHARACTERISTICS

INPUT OFFSET VOLTAGE OF REPRESENTATIVE INDIVIDUAL UNITS vs FREE-AIR TEMPERATURE

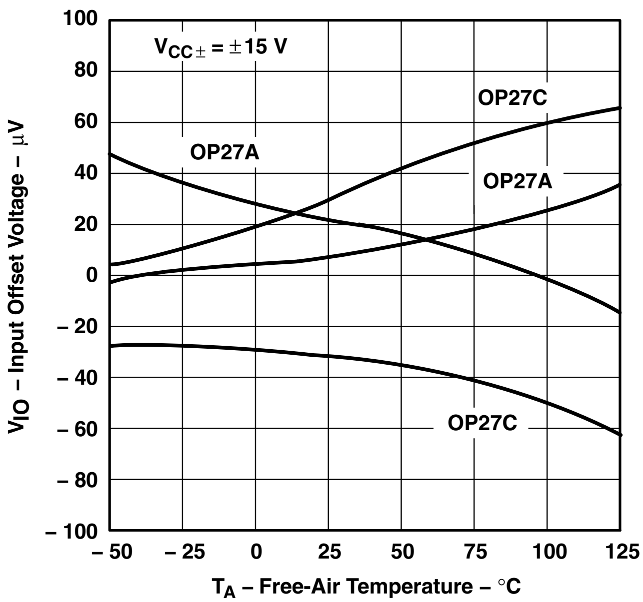


Figure 1

WARM-UP CHANGE IN INPUT OFFSET VOLTAGE vs ELAPSED TIME

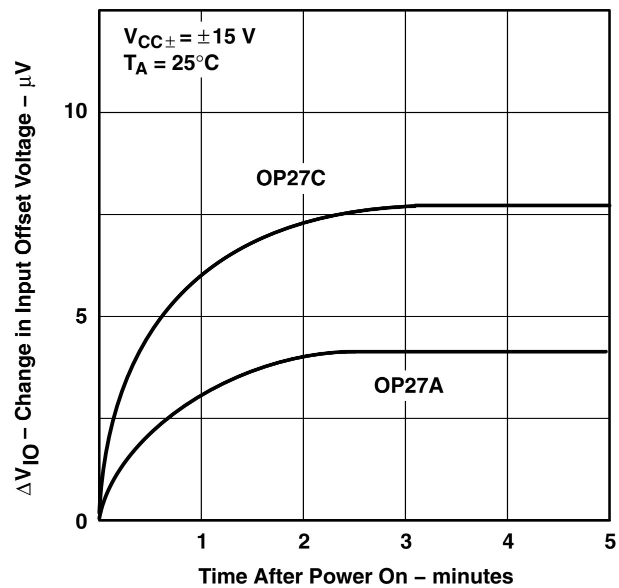


Figure 2

LONG-TERM DRIFT OF INPUT OFFSET VOLTAGE OF REPRESENTATIVE INDIVIDUAL UNITS

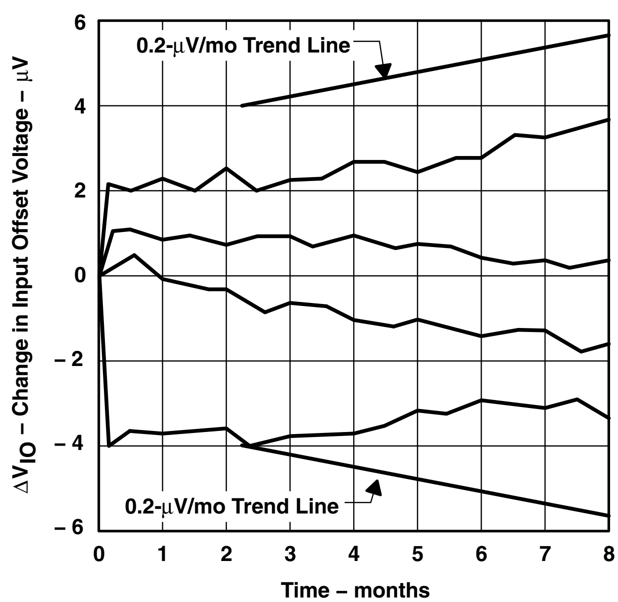


Figure 3

TYPICAL CHARACTERISTICS

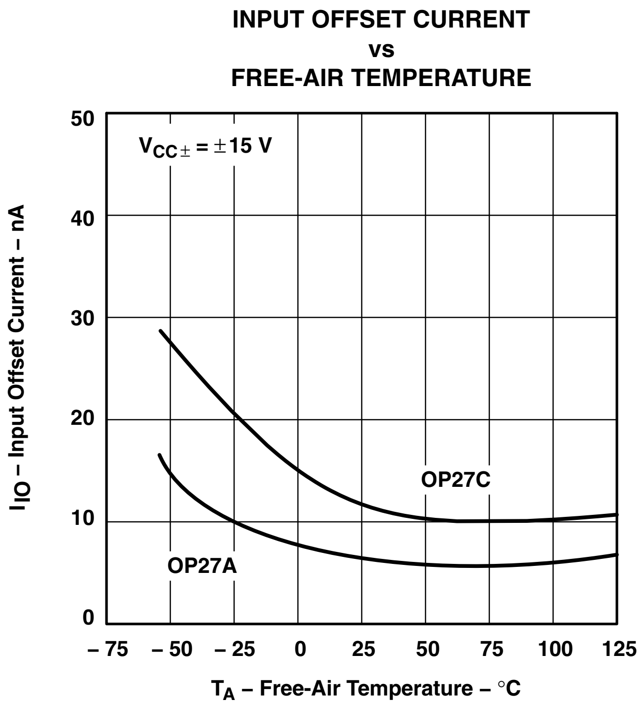


Figure 4

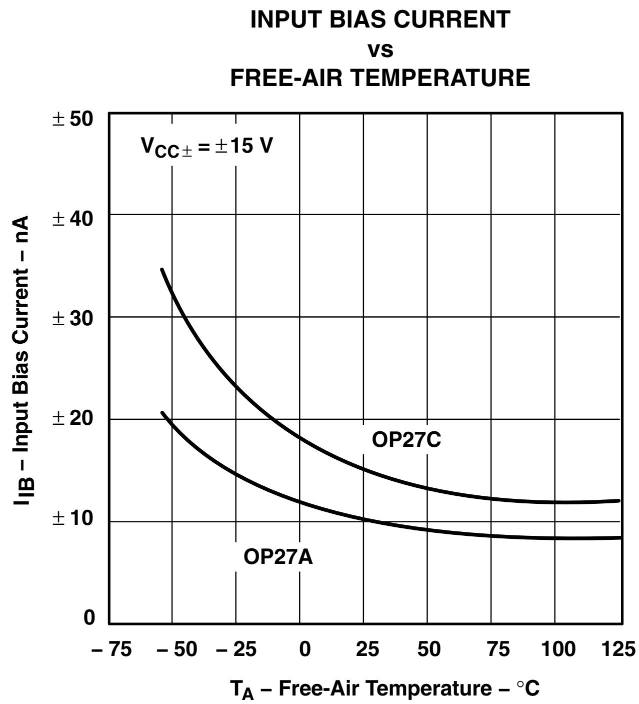


Figure 5

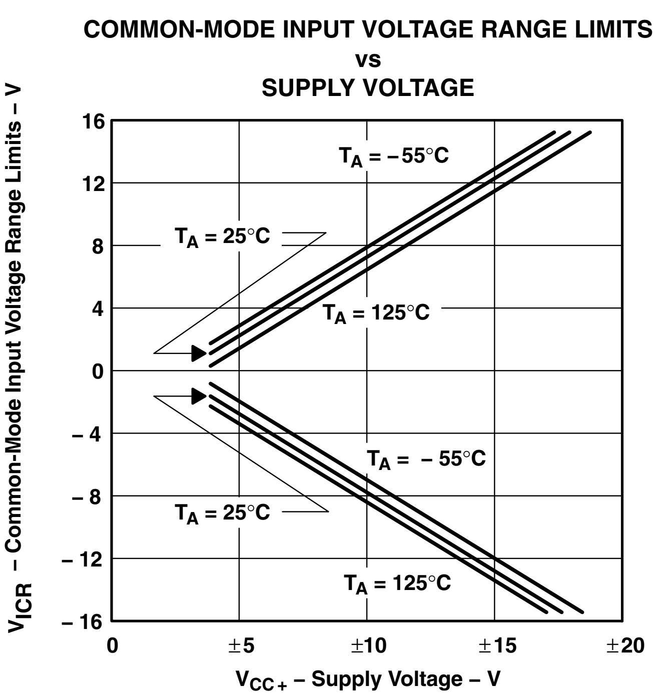


Figure 6

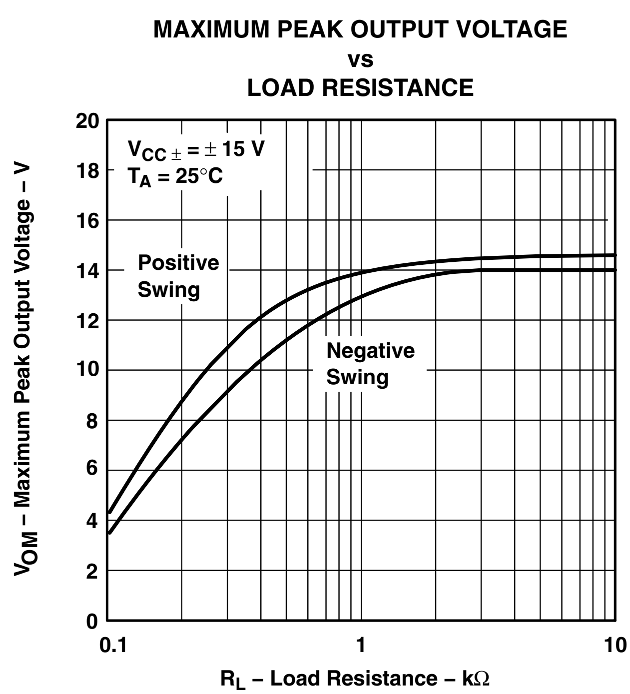


Figure 7

TYPICAL CHARACTERISTICS

OP27
MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE vs FREQUENCY

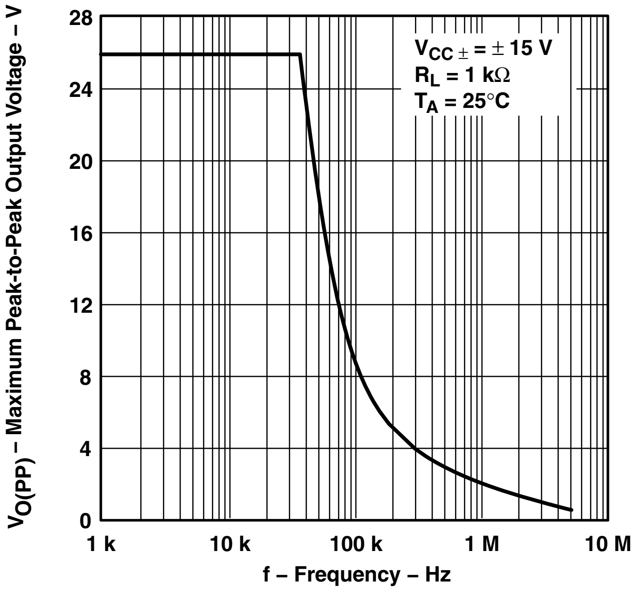


Figure 8.

OP27A
LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs TOTAL SUPPLY VOLTAGE

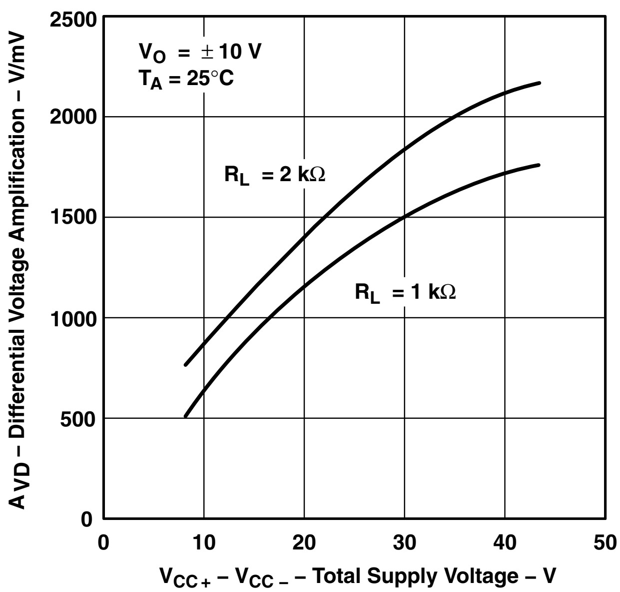


Figure 9

OP27A
LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs LOAD RESISTANCE

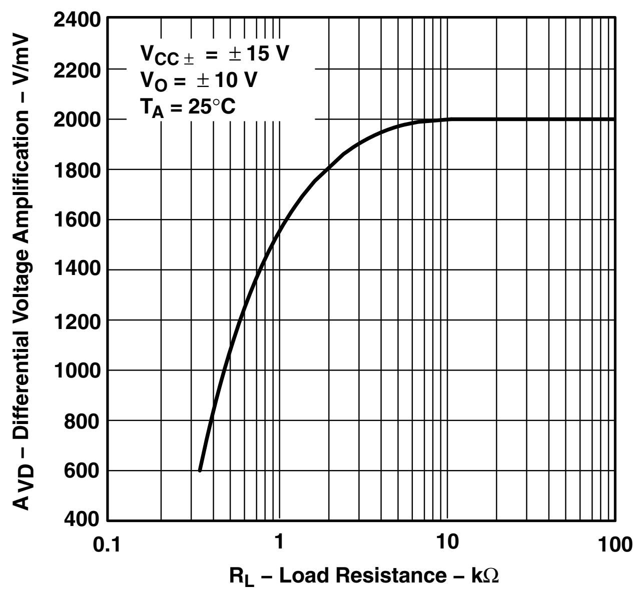


Figure 10

TYPICAL CHARACTERISTICS

**OP27
LARGE-SIGNAL DIFFERENTIAL
VOLTAGE AMPLIFICATION AND PHASE SHIFT
vs
FREQUENCY**

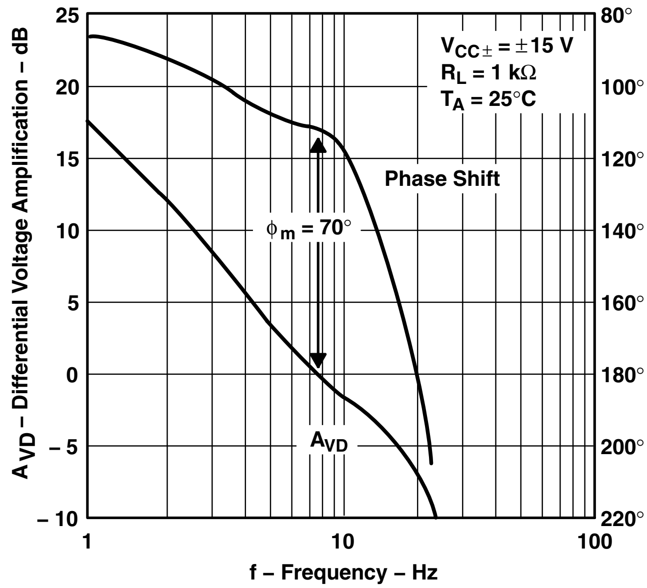


Figure 11.

**OP27A
LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREQUENCY**

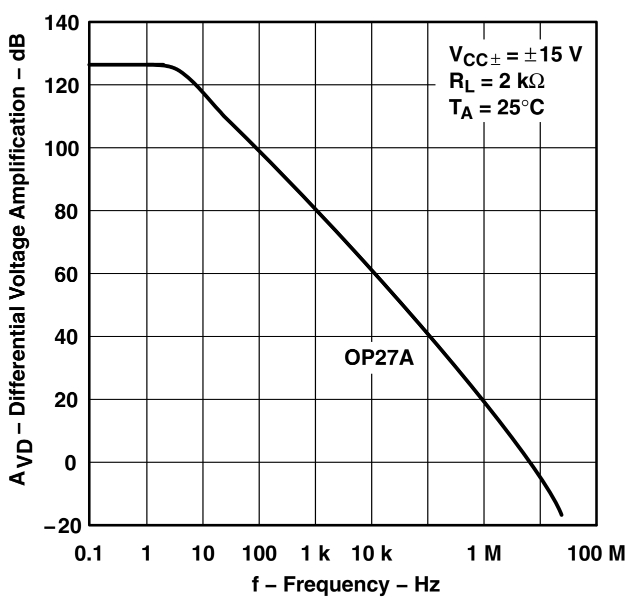


Figure 12

**OP27A
COMMON-MODE REJECTION RATIO
vs
FREQUENCY**

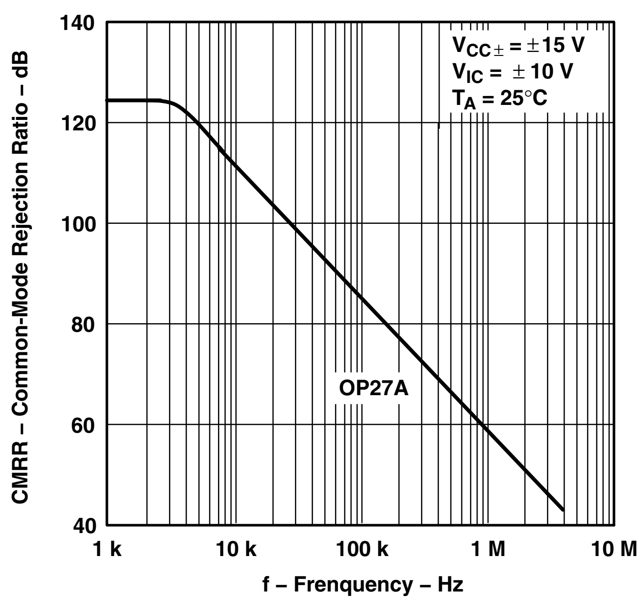


Figure 13

TYPICAL CHARACTERISTICS

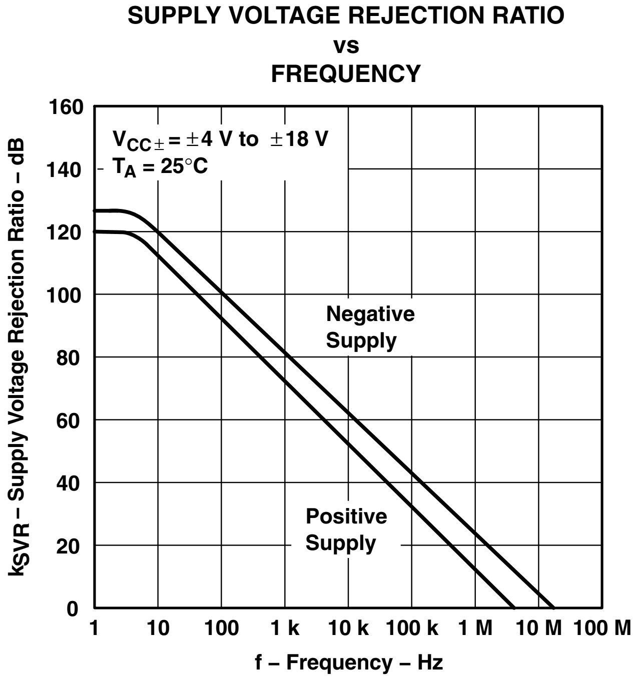


Figure 14

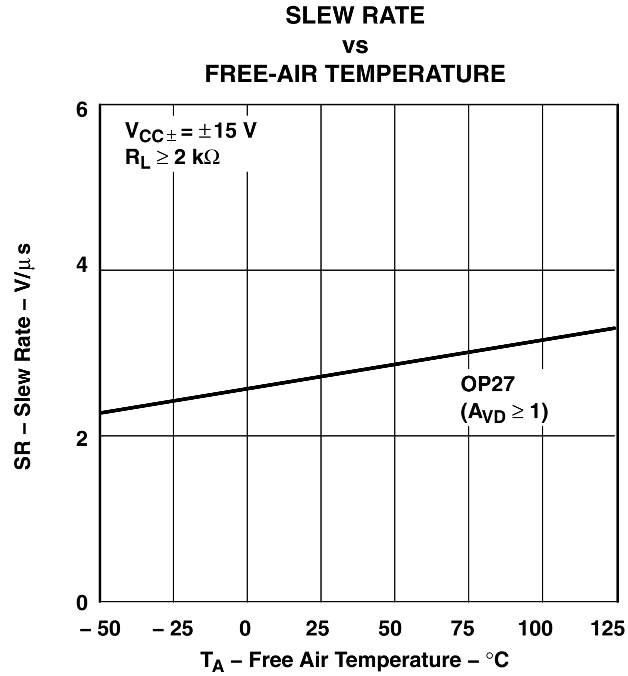


Figure 15

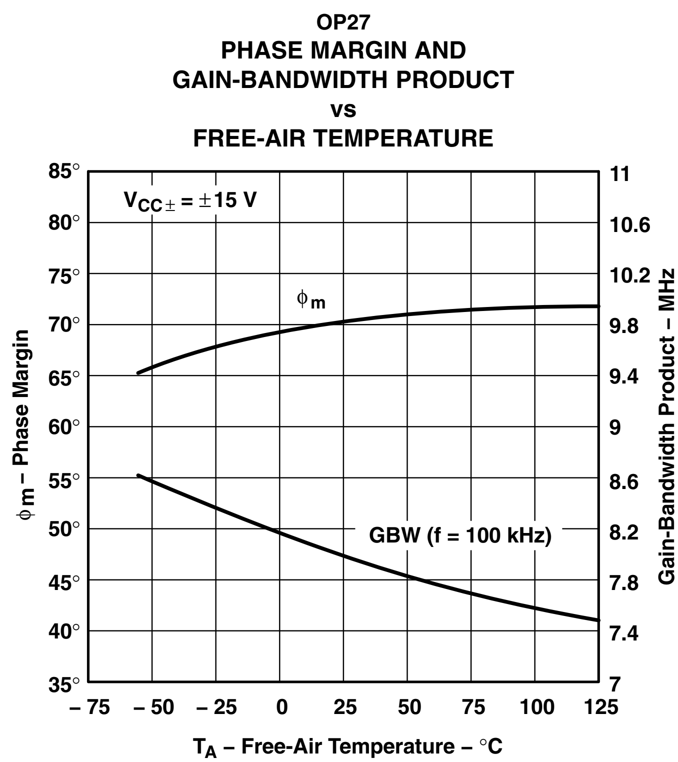


Figure 16.

TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE vs BANDWIDTH

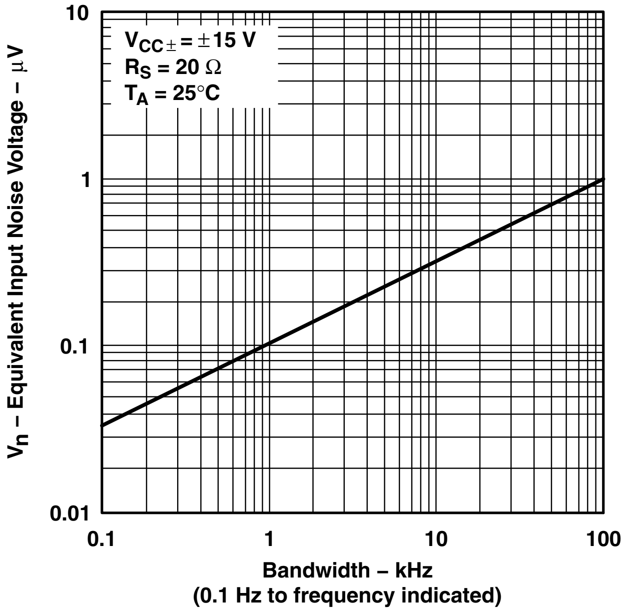


Figure 17

TOTAL EQUIVALENT INPUT NOISE VOLTAGE vs SOURCE RESISTANCE

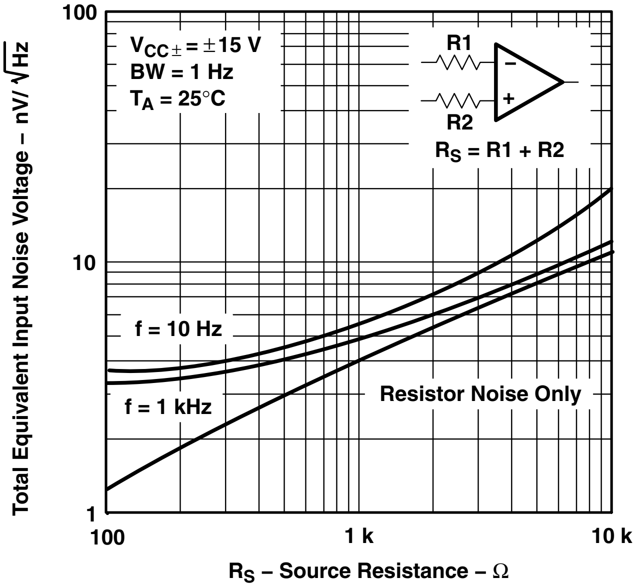


Figure 18

OP27A EQUIVALENT INPUT NOISE VOLTAGE vs TOTAL SUPPLY VOLTAGE

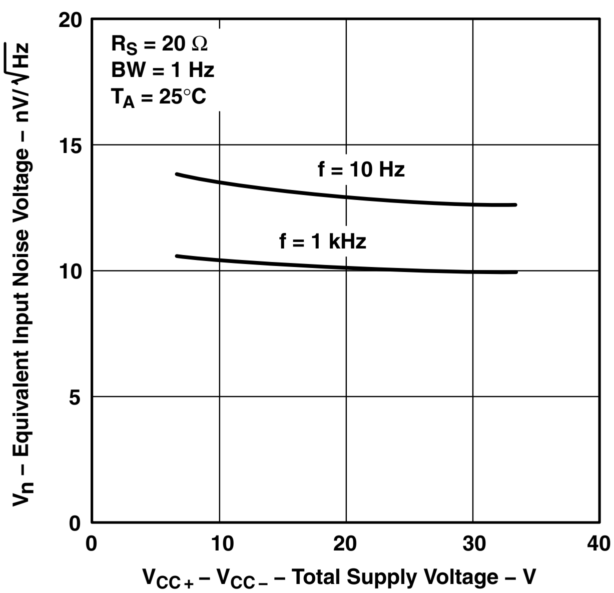


Figure 19

OP27A EQUIVALENT INPUT NOISE VOLTAGE vs FREE-AIR TEMPERATURE

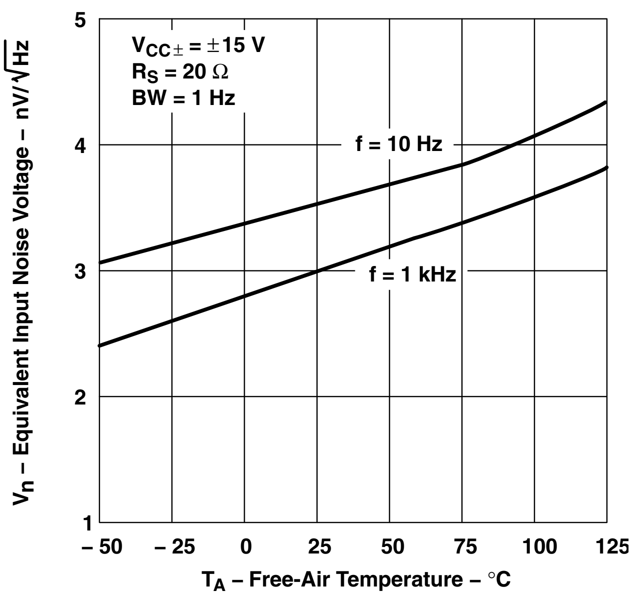


Figure 20

TYPICAL CHARACTERISTICS

OP27A
EQUIVALENT INPUT NOISE VOLTAGE
vs
FREQUENCY

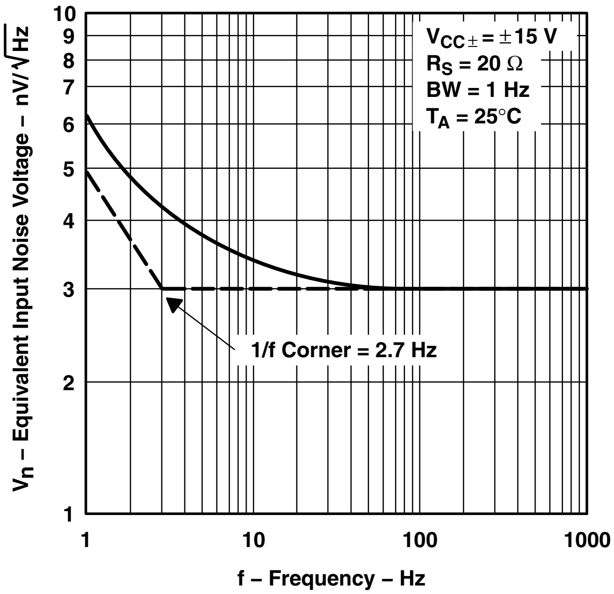


Figure 21

SHORT-CIRCUIT OUTPUT CURRENT
vs
ELAPSED TIME

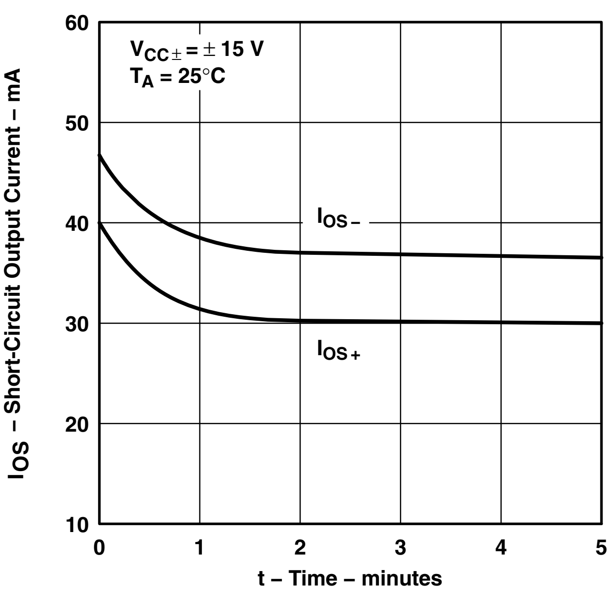


Figure 22

SUPPLY CURRENT
vs
TOTAL SUPPLY VOLTAGE

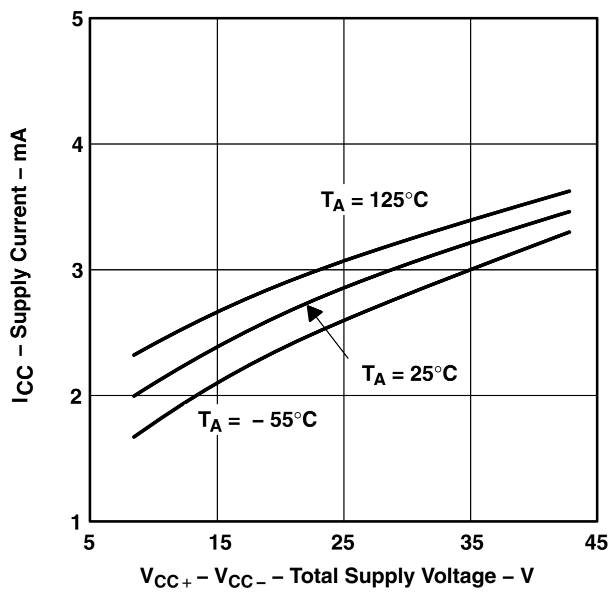


Figure 23

TYPICAL CHARACTERISTICS

**OP27
VOLTAGE FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE**

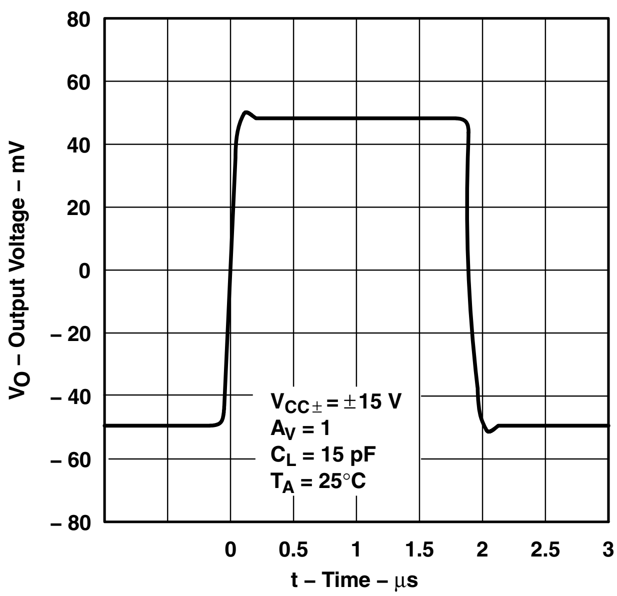


Figure 24

**OP27
VOLTAGE FOLLOWER
LARGE-SIGNAL
PULSE RESPONSE**

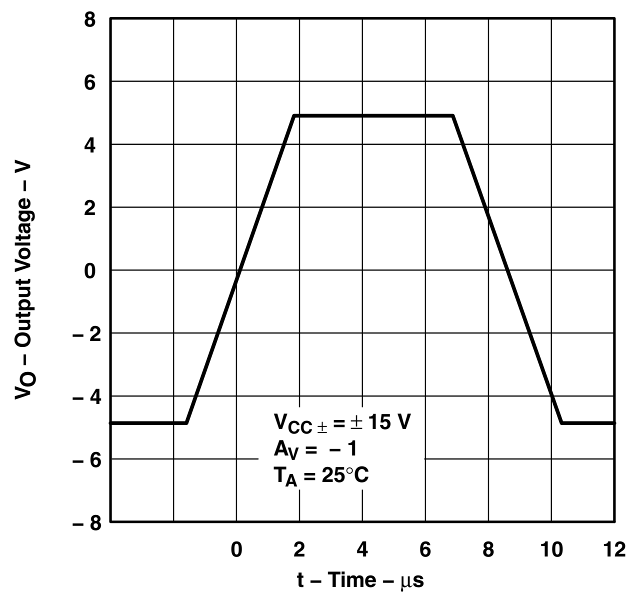


Figure 25

APPLICATION INFORMATION

general

The OP27 series devices can be inserted directly onto OP07, OP05, µA725, and SE5534 sockets with or without removing external compensation or nulling components. In addition, the OP27 can be fitted to µA741 sockets by removing or modifying external nulling components.

noise testing

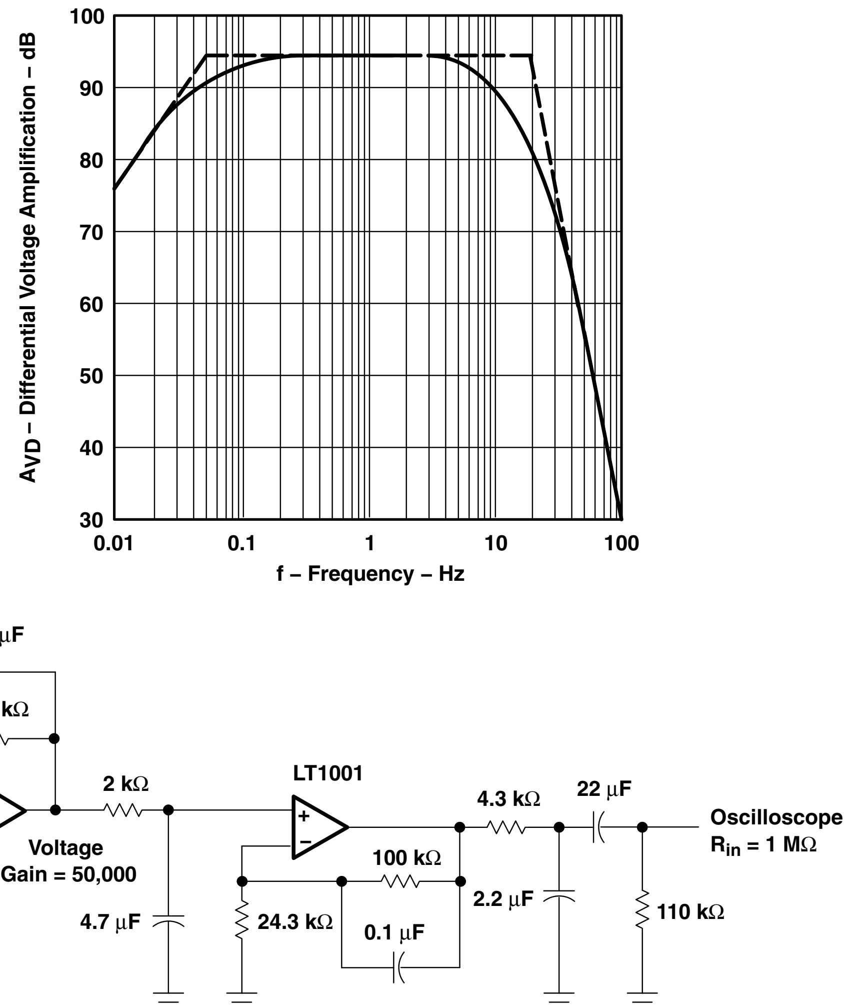
Figure 26 shows a test circuit for 0.1-Hz to 10-Hz peak-to-peak noise measurement of the OP27. The frequency response of this noise tester indicates that the 0.1-Hz corner is defined by only one zero. Because the time limit acts as an additional zero to eliminate noise contributions from the frequency band below 0.1 Hz, the test time to measure 0.1-Hz to 10-Hz noise should not exceed 10 seconds.

Measuring the typical 80-nV peak-to-peak noise performance of the OP27 requires the following special test precautions:

APPLICATION INFORMATION

noise testing (continued)

1. The device should be warmed up for at least five minutes. As the operational amplifier warms up, the offset voltage typically changes 4 μV due to the chip temperature increasing from 10°C to 20°C starting from the moment the power supplies are turned on. In the 10-s measurement interval, these temperature-induced effects can easily exceed tens of nanovolts.
2. For similar reasons, the device should be well shielded from air currents to eliminate the possibility of thermoelectric effects in excess of a few nanovolts, which would invalidate the measurements.
3. Sudden motion in the vicinity of the device should be avoided, as it produces a feedthrough effect that increases observed noise.



NOTE: All capacitor values are for nonpolarized capacitors only.

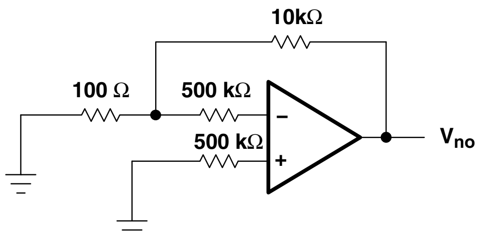
Figure 26. 0.1-Hz to 10-Hz Peak-to-Peak Noise Test Circuit and Frequency Response

APPLICATION INFORMATION

noise testing (continued)

When measuring noise on a large number of units, a noise-voltage density test is recommended. A 10-Hz noise-voltage density measurement correlates well with a 0.1-Hz to 10-Hz peak-to-peak noise reading since both results are determined by the white noise and the location of the 1/f corner frequency.

Figure 27 shows a circuit measuring current noise and the formula for calculating current noise.



$$I_n = \frac{[V_{no}^2 - (130\ nV)^2]^{1/2}}{1\ M\Omega \times 100}$$

Figure 27. Current Noise Test Circuit and Formula

offset voltage adjustment

The input offset voltage and temperature coefficient of the OP27 are permanently trimmed to a low level at wafer testing. However, if further adjustment of V_{IO} is necessary, using a 10-kΩ nulling potentiometer as shown in Figure 28 does not degrade the temperature coefficient α_{VIO}. Trimming to a value other than zero creates an α_{VIO} of V_{IO}/300 µV/°C. For example, if V_{IO} is adjusted to 300 µV, the change in α_{VIO} is 1 µV/°C.

The adjustment range with a 10-kΩ potentiometer is approximately ±2.5 mV. If a smaller adjustment range is needed, the sensitivity and resolution of the nulling can be improved by using a smaller potentiometer in conjunction with fixed resistors. The example in Figure 29 has an approximate null range of ±200 µV.

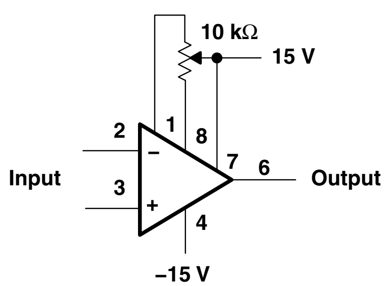


Figure 28. Standard Input Offset Voltage Adjustment

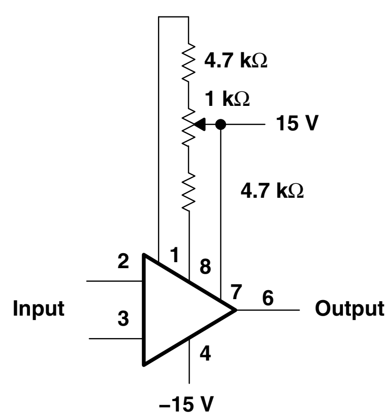


Figure 29. Input Offset Voltage Adjustment With Improved Sensitivity

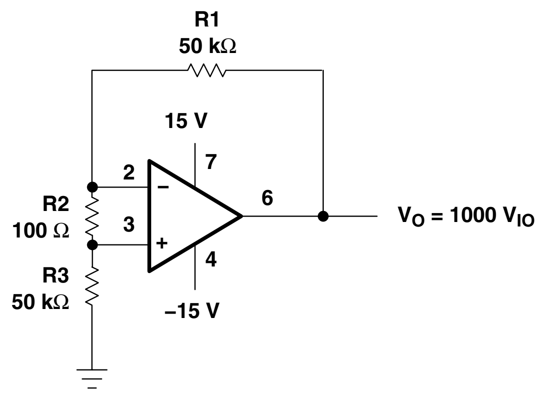
offset voltage and drift

Unless proper care is exercised, thermoelectric effects caused by temperature gradients across dissimilar metals at the contacts to the input terminals can exceed the inherent temperature coefficient $\propto V_{IO}$ of the amplifier. Air currents should be minimized, package leads should be short, and the two input leads should be close together and at the same temperature.

APPLICATION INFORMATION

offset voltage and drift (continued)

The circuit shown in Figure 30 measures offset voltage. This circuit can also be used as the burn-in configuration for the OP27 with the supply voltage increased to 20 V, R1 = R3 = 10 kΩ, R2 = 200 Ω, and A_{VD} = 100.



NOTE A: Resistors must have low thermoelectric potential.

Figure 30. Test Circuit for Offset Voltage and Offset Voltage Temperature Coefficient

unity gain buffer applications

The resulting output waveform, when R_f ≤ 100 Ω and the input is driven with a fast large-signal pulse (>1 V), is shown in the pulsed-operation diagram in Figure 31.

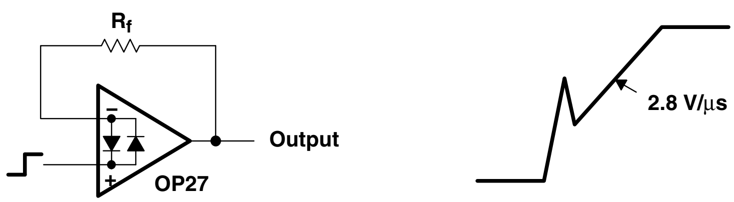
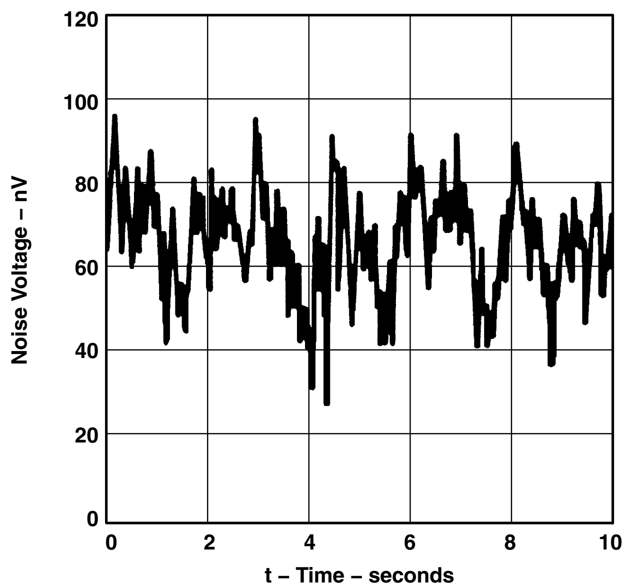


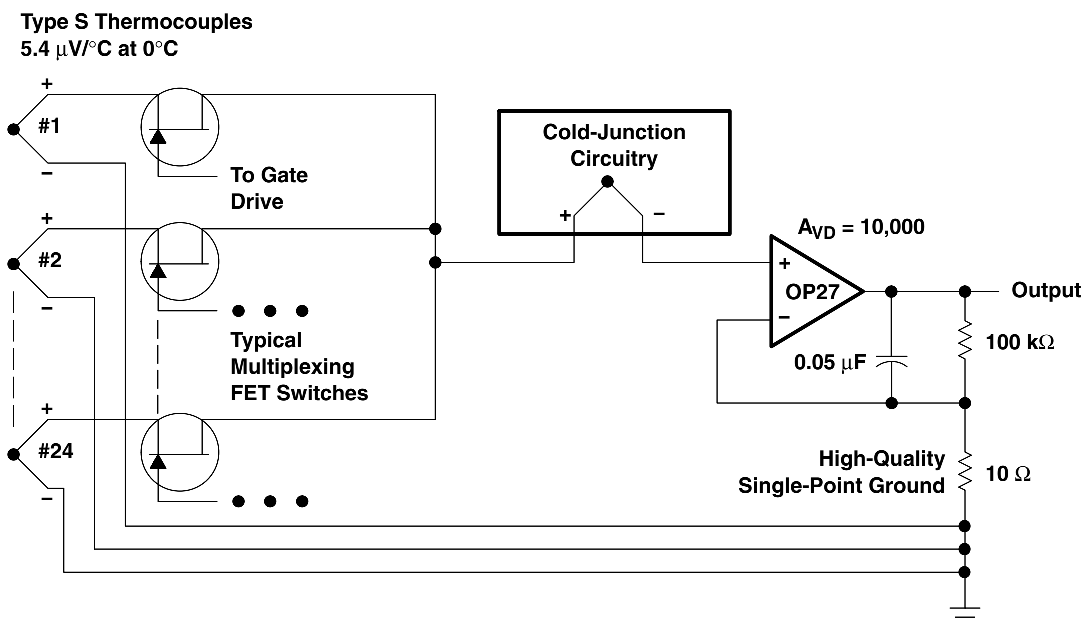
Figure 31. Pulsed Operation

During the initial (fast-feedthrough-like) portion of the output waveform, the input protection diodes effectively short the output to the input, and a current, limited only by the output short-circuit protection, is drawn by the signal generator. When R_f ≥ 500 Ω, the output is capable of handling the current requirements (load current ≤20 mA at 10 V), the amplifier stays in its active mode, and a smooth transition occurs. When R_f > 2 kΩ, a pole is created with R_f and the amplifier's input capacitance, creating additional phase shift and reducing the phase margin. A small capacitor (20 pF to 50 pF) in parallel with R_f eliminates this problem.

APPLICATION INFORMATION

unity gain buffer applications (continued)





NOTE A: If 24 channels are multiplexed per second and the output is required to settle to 0.1 % accuracy, the amplifier's bandwidth cannot be limited to less than 30 Hz. The peak-to-peak noise contribution of the OP27 will still be only 0.11 μV, which is equivalent to an error of only 0.02°C.

Figure 32. Low-Noise, Multiplexed Thermocouple Amplifier and 0.1-Hz to 10-Hz Peak-to-Peak Noise Voltage

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/ Ball Finish	MSL Peak Temp (3)	Samples (Requires Login)
JM38510/13503BPA	OBSOLETE	CDIP	JG	8		TBD	Call TI	Call TI	Replaced by JM38510/13506BPA
JM38510/13506BPA	ACTIVE	CDIP	JG	8	1	TBD	A42	N / A for Pkg Type	Purchase Samples
OP27AFKB	ACTIVE	LCCC	FK	20	1	TBD	POST-PLATE	N / A for Pkg Type	Purchase Samples
OP27AJGB	ACTIVE	CDIP	JG	8	1	TBD	A42	N / A for Pkg Type	Purchase Samples
OP27CJGB	ACTIVE	CDIP	JG	8	1	TBD	A42	N / A for Pkg Type	Purchase Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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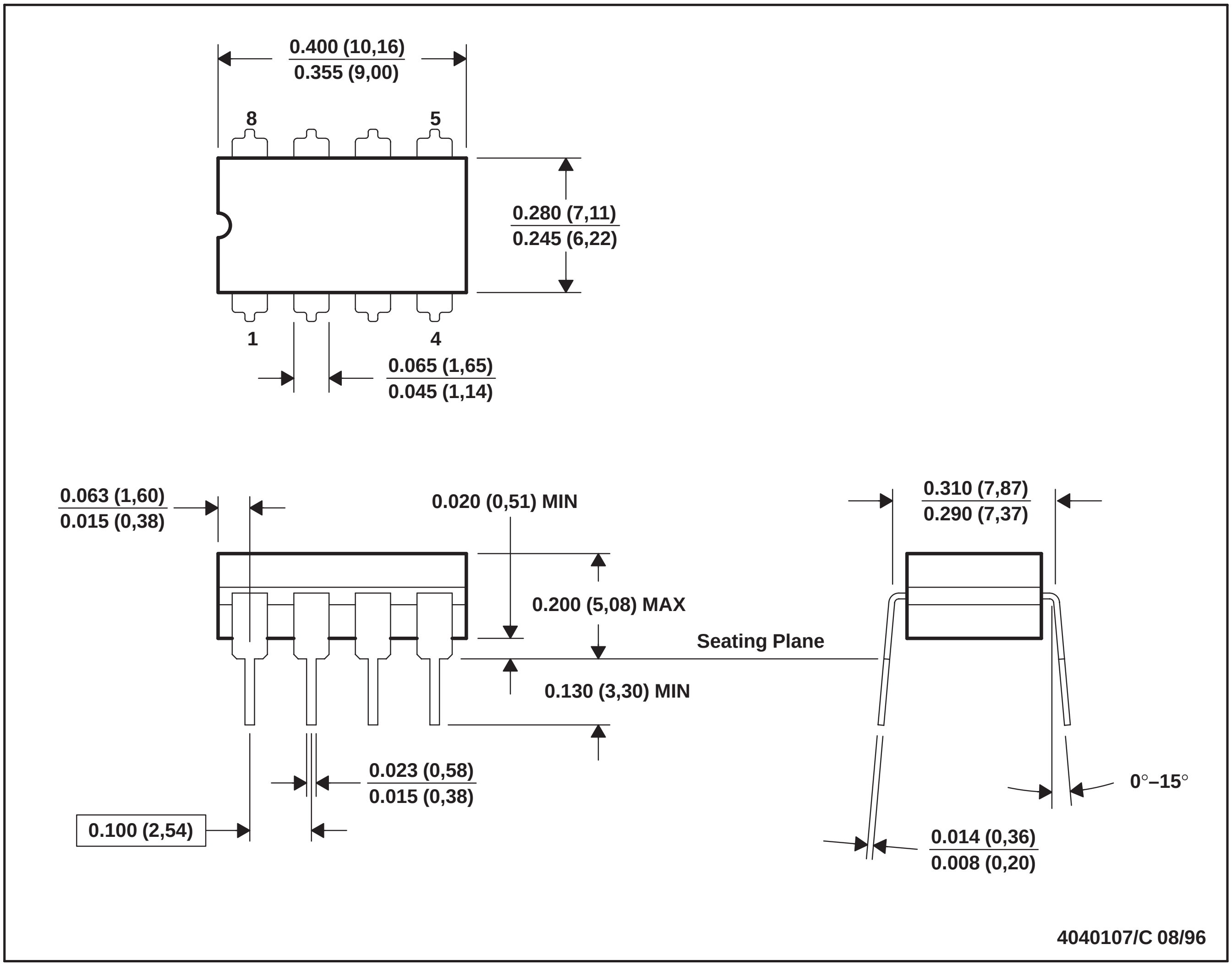
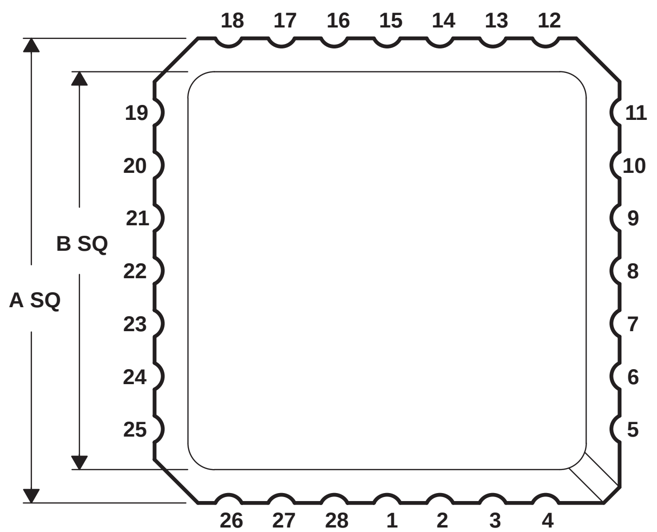
JG (R-GDIP-T8) **CERAMIC DUAL-IN-LINE**



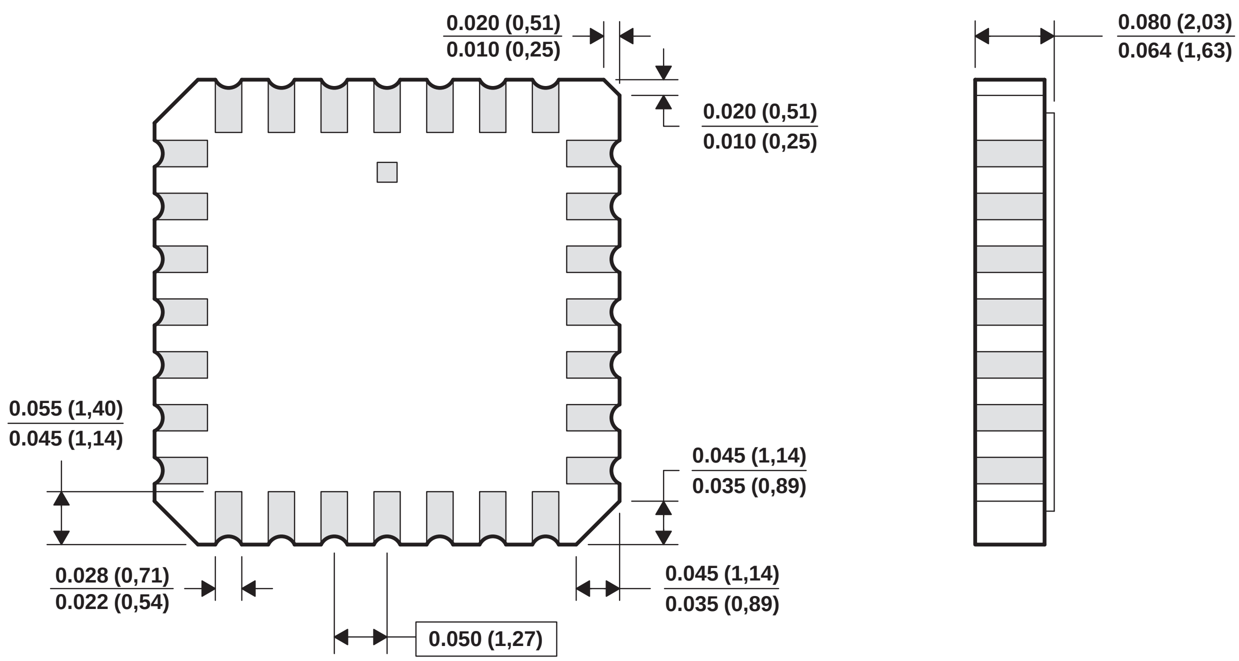
NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. This package can be hermetically sealed with a ceramic lid using glass frit.
D. Index point is provided on cap for terminal identification.
E. Falls within MIL STD 1835 GDIP1-T8

FK (S-CQCC-N)** **LEADLESS CERAMIC CHIP CARRIER**

28 TERMINAL SHOWN



NO. OF TERMINALS **	A		B	
	MIN	MAX	MIN	MAX
20	0.342 (8,69)	0.358 (9,09)	0.307 (7,80)	0.358 (9,09)
28	0.442 (11,23)	0.458 (11,63)	0.406 (10,31)	0.458 (11,63)
44	0.640 (16,26)	0.660 (16,76)	0.495 (12,58)	0.560 (14,22)
52	0.739 (18,78)	0.761 (19,32)	0.495 (12,58)	0.560 (14,22)
68	0.938 (23,83)	0.962 (24,43)	0.850 (21,6)	0.858 (21,8)
84	1.141 (28,99)	1.165 (29,59)	1.047 (26,6)	1.063 (27,0)



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NOTES: A. All linear dimensions are in inches (millimeters).
B. This drawing is subject to change without notice.
C. This package can be hermetically sealed with a metal lid.
D. The terminals are gold plated.
E. Falls within JEDEC MS-004

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DLP® Products	www.dlp.com	Communications and Telecom	www.ti.com/communications
DSP	dsp.ti.com	Computers and Peripherals	www.ti.com/computers
Clocks and Timers	www.ti.com/clocks	Consumer Electronics	www.ti.com/consumer-apps
Interface	interface.ti.com	Energy	www.ti.com/energy
Logic	logic.ti.com	Industrial	www.ti.com/industrial
Power Mgmt	power.ti.com	Medical	www.ti.com/medical
Microcontrollers	microcontroller.ti.com	Security	www.ti.com/security
RFID	www.ti-rfid.com	Space, Avionics & Defense	www.ti.com/space-avionics-defense
RF/IF and ZigBee® Solutions	www.ti.com/lprf	Video and Imaging	www.ti.com/video
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